

Theory and implementation of a precision lock-in frequency meter

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- 1、Background
- 2、Theory and simulation
- 3、Implementation results
- 4、Summary

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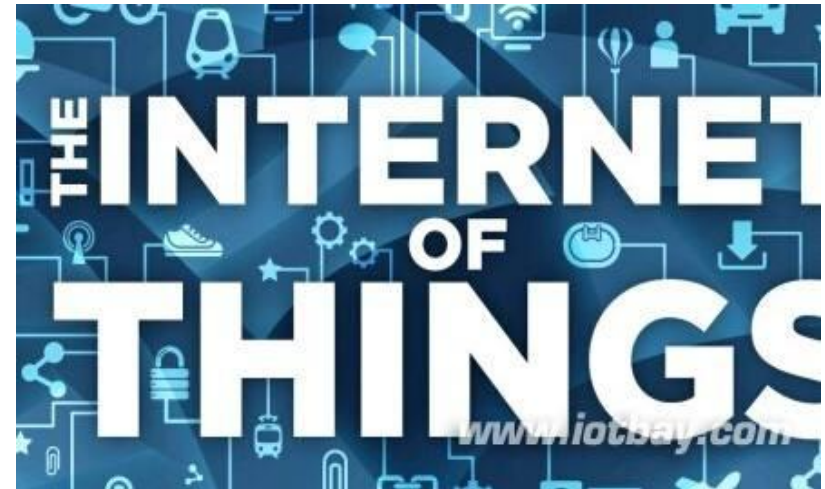
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I.I why precision frequency measurement?

Better GPS Positioning



Higher 5G capacitance



Ultra precise scientific experiments



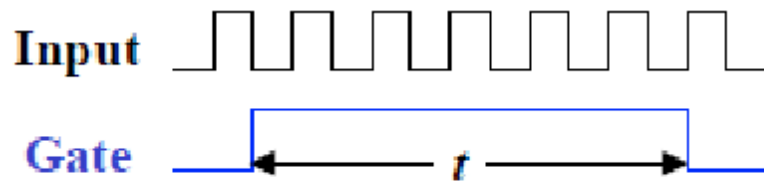
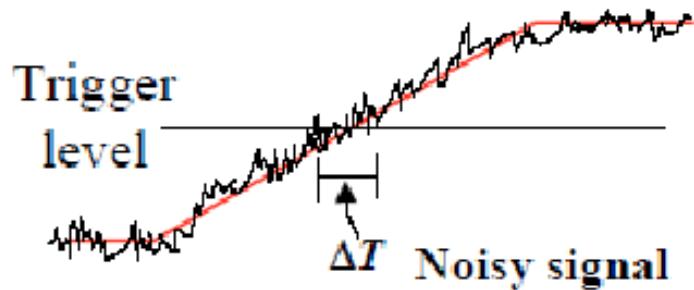
Finer control and fabrication



1.2 Long history of frequency measurement

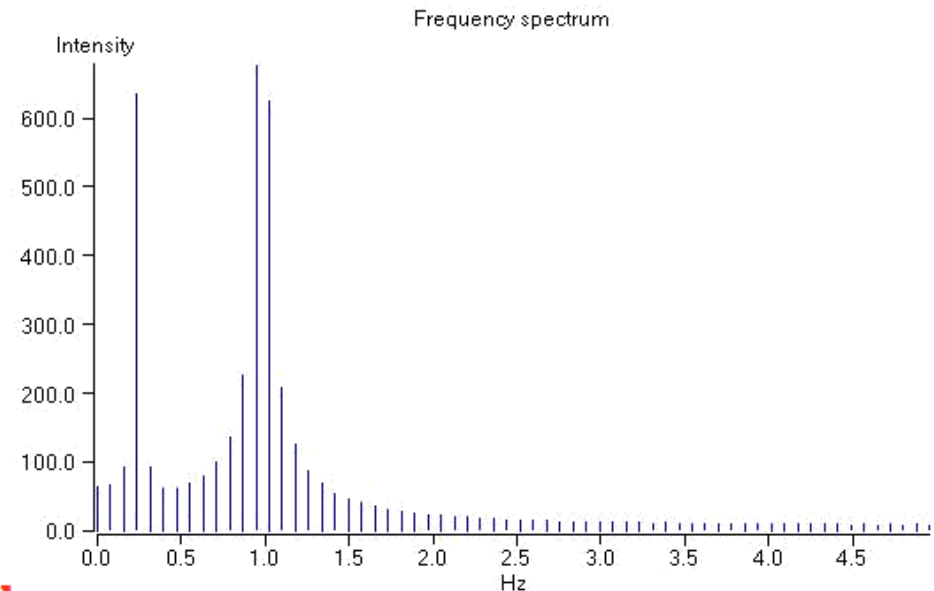
$$f = \frac{N}{t}$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$



N pulses

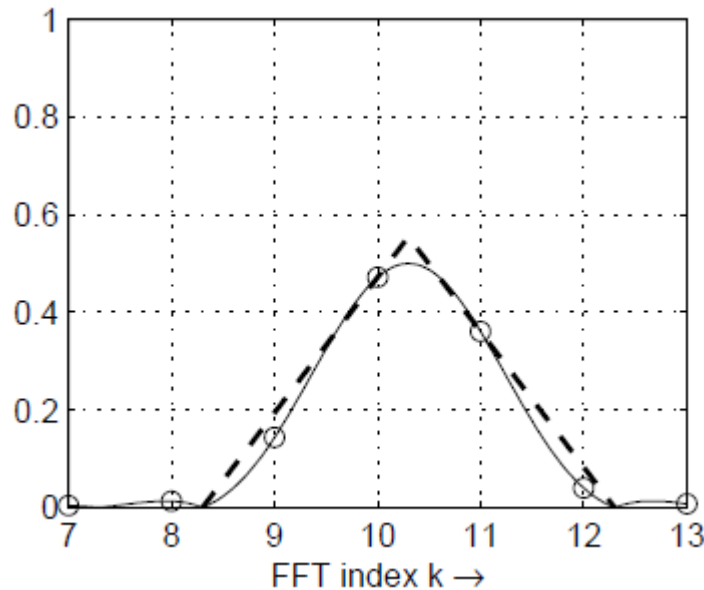
In time domain: to counter
and measure duration.
(Only for high SNR)



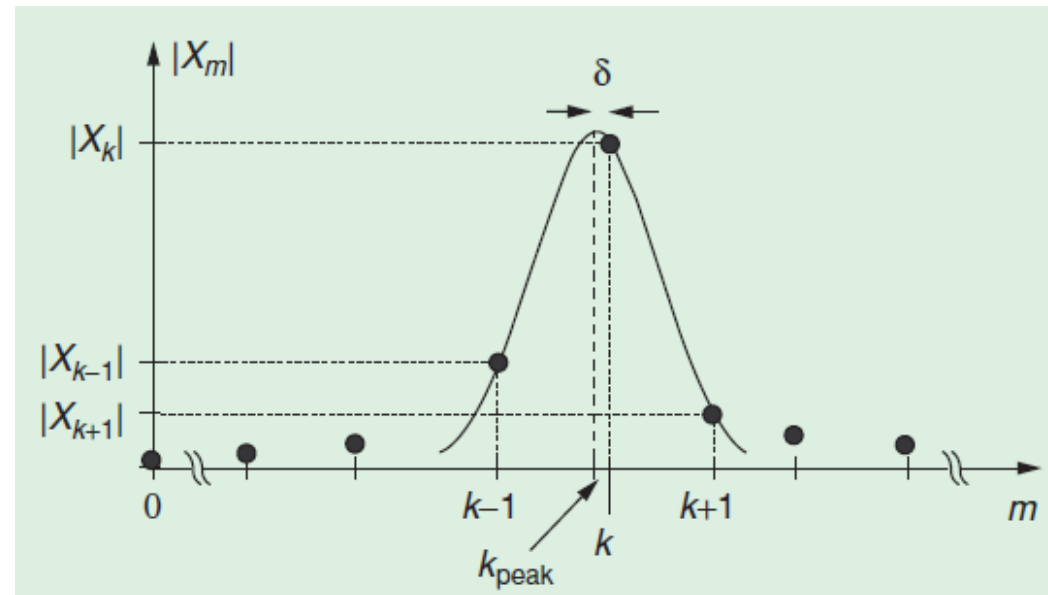
In frequency domain:
Fourier transformation.
(Only for coarse estimation)

I.3 To push the precision: previous work(1/3)

Triangle function fitting the 1st order approximation



Parabolic function fitting: the 2nd order approximation

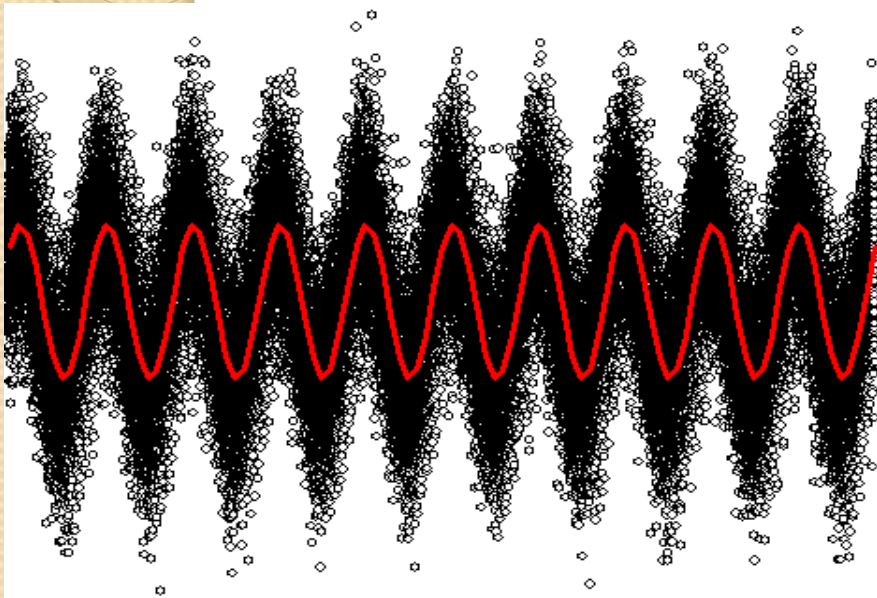


Eric Jacobsen and Peter Kootsookos, Fast accurate frequency estimators, IEEE Sig. Proc. Mag., 2007, 123(2007)

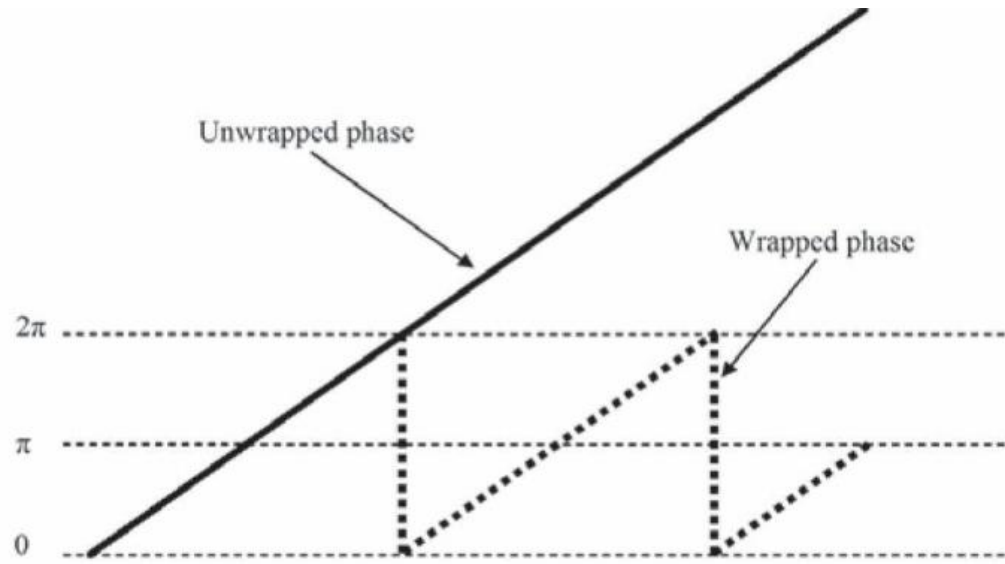
K. Hofbauer, Estimating frequency and amplitude of sinusoids a survey and the use of shifted Fourier transforms, Diploma Thesis, Graz Univ. Tech., 2004

I.4 To push the precision: previous work(2/3)

Maximum likelihood estimation



Phase unwrapping



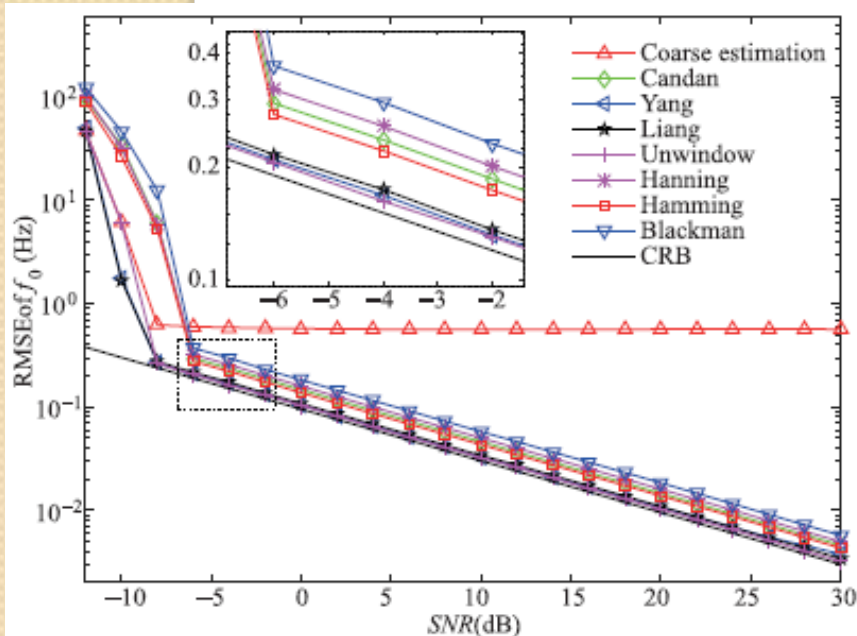
Z. Shen and S. Zhang, Improved frequency estimation algorithm by least squares phase unwrapping, *Circuits Syst. Signal Proc.*, 37, 5680-7(2018)

D.C. Rife and R.R. Boorstyn, Multiple tone parameter estimation from discrete-time observations, *Bell Sys. Tech. J.*, 55(9), 1389-410(1976)

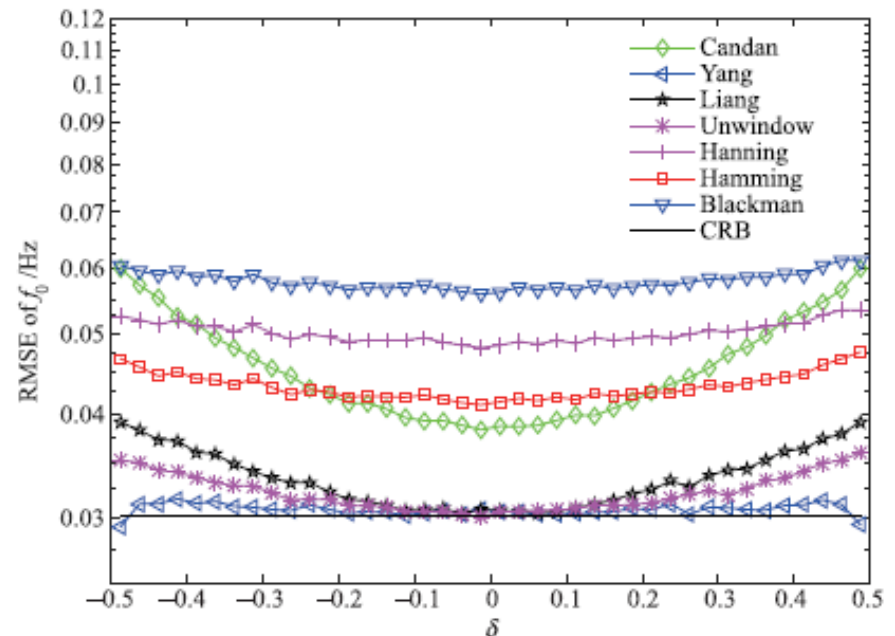
I.5 To push the precision: previous work(3/3)

Cramer-Raw Lower Bound

$$\text{var}(\omega) = \frac{12\sigma^2}{b_0^2 T^2 N(N^2 - 1)}$$



Windows optimization



Unfortunately, **getting frequency at low SNR** with proper results, still remains a challenge.

J.Z. Xiang, W. Cui and Q. Shen, Chin. J. Electr., 27(1), 109-14(2018)

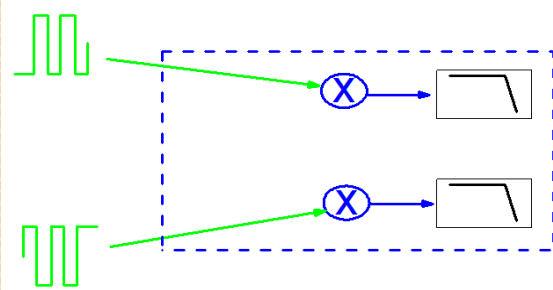
D.C. Rife and R.R. Boorstyn, IEEE Trans. Inform. Theor., 20, 591-8(1974)

2.1 Concepts of the lock-in amplifiers(LIAs)

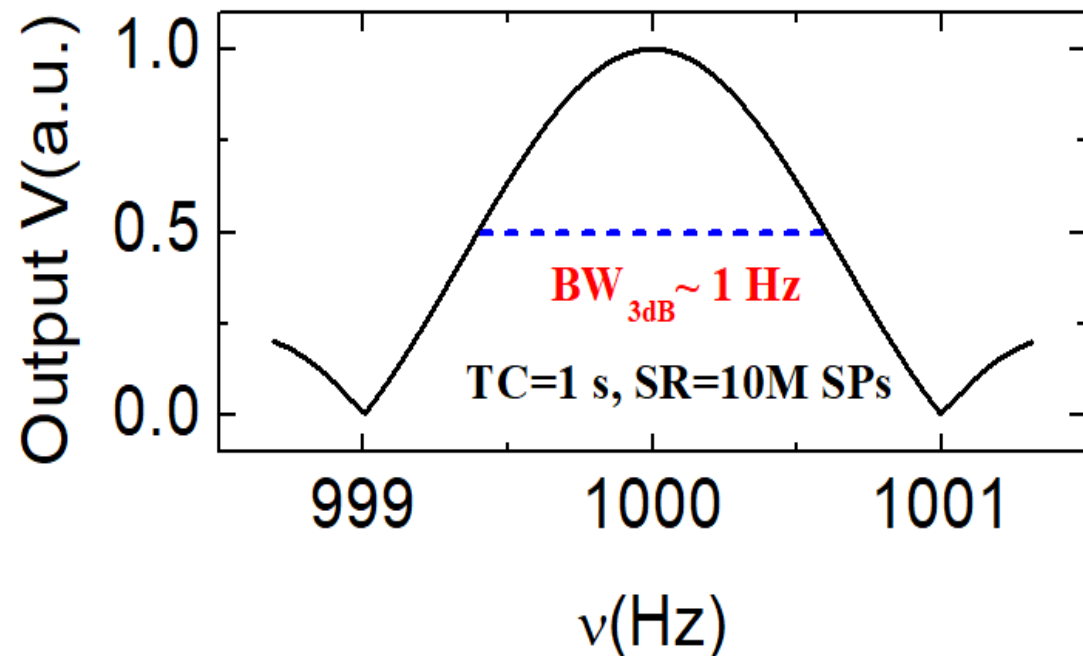
In principle, LIAs work as **homodyning** mixers for phase sensitive detection(**PSD**)

As an **ultra narrow** band pass filter

$$S_{\text{Re}}(\omega) = \frac{2}{T} \int_{t=0}^T [S(t) \times \cos(\omega t)] dt$$



$$S_{\text{Im}}(\omega) = \frac{2}{T} \int_{t=0}^T [S(t) \times \sin(\omega t)] dt$$

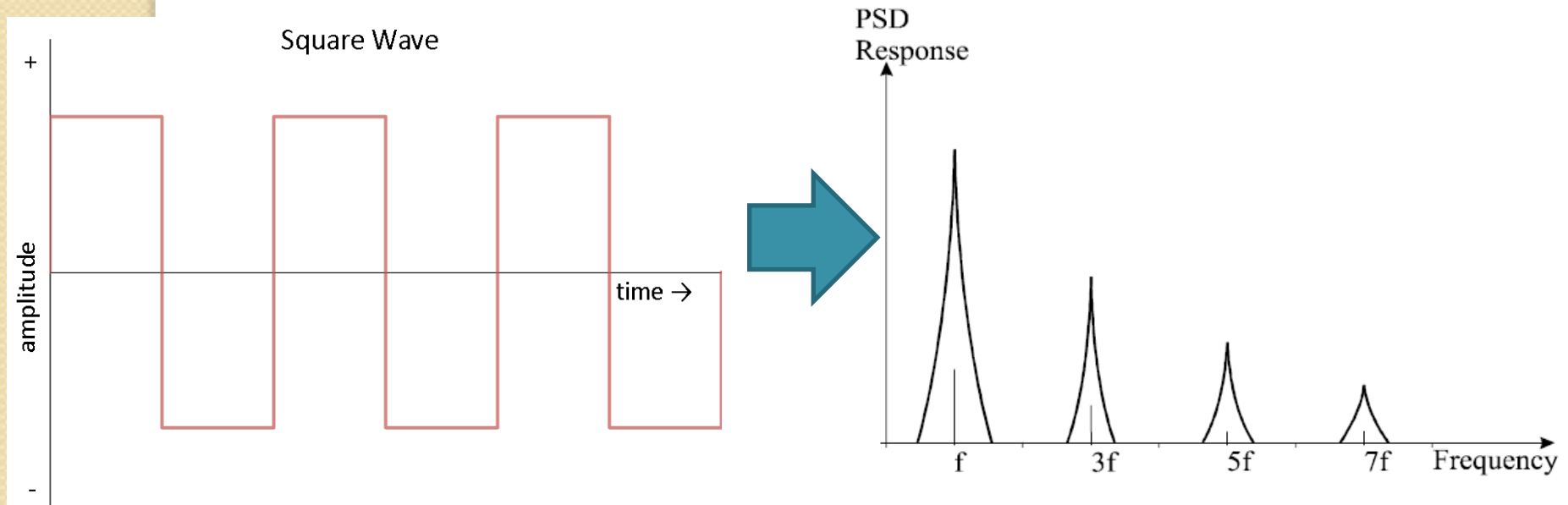


LIAs can easily separate in-phase and quadrature phase components

M. L. Meade, Lock-In Amplifiers: Principles and Applications. (Peter Peregrinus, London, 1983)

2.2 Classic LIAs work at fixed frequencies

Single reference input is normally required.



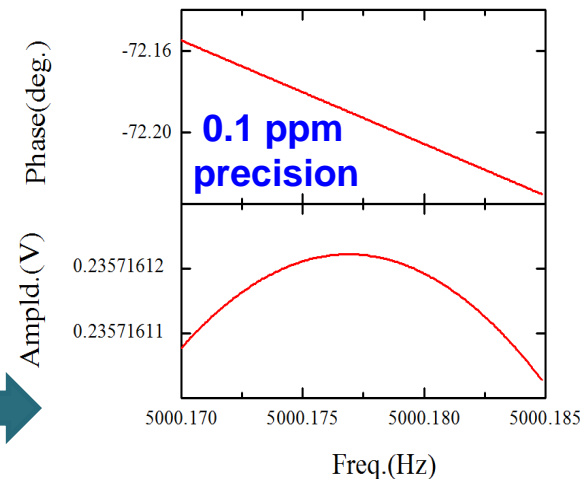
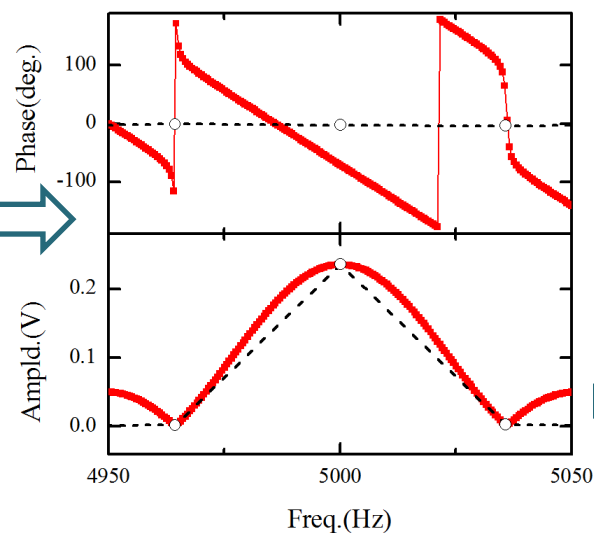
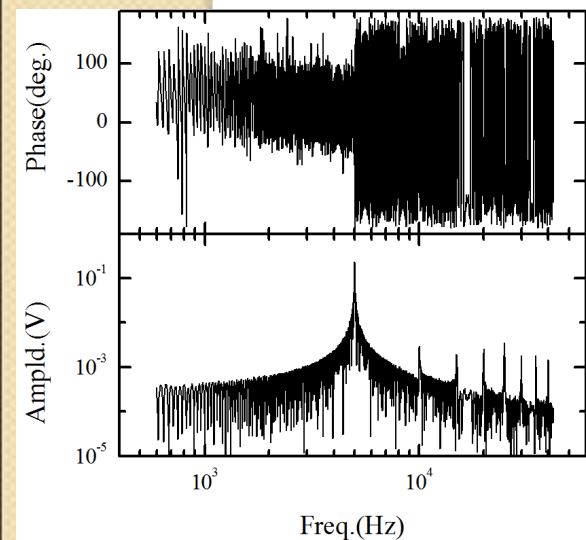
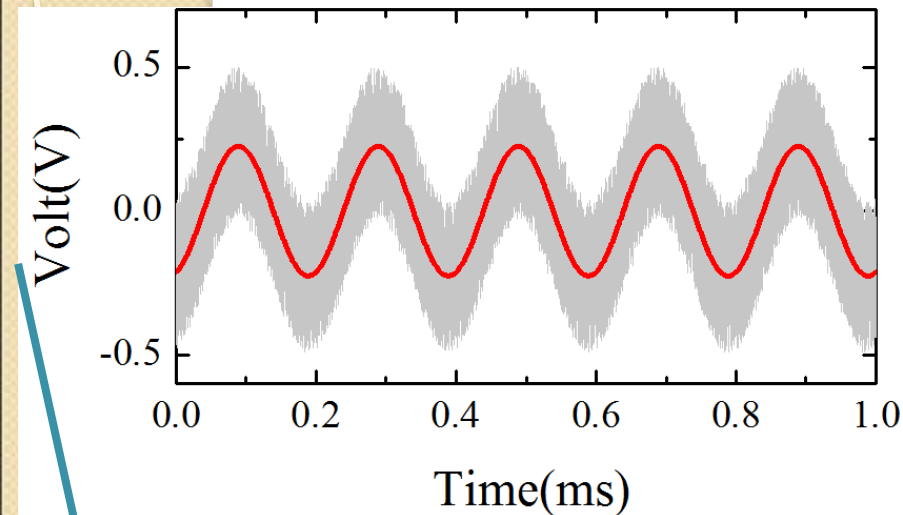
With switches, other frequencies such as harmonic ones could also be measured(**heterodyning mode**).

2.3 Lu's Fine scanning LIAs: lock-in frequencymeter LIFs

Based on **correlation functional theory**

$$R_{ij} = \frac{1}{N} \sum_{k=0}^N [S_k \times S_{\text{ref},k}(f_i, \Psi_j, 1)]$$

FFT could not finely tuned in frequency range.



Frequency taken as a variable, and no reference input needed.

2.4 Theory on the local spectrum of LIFs

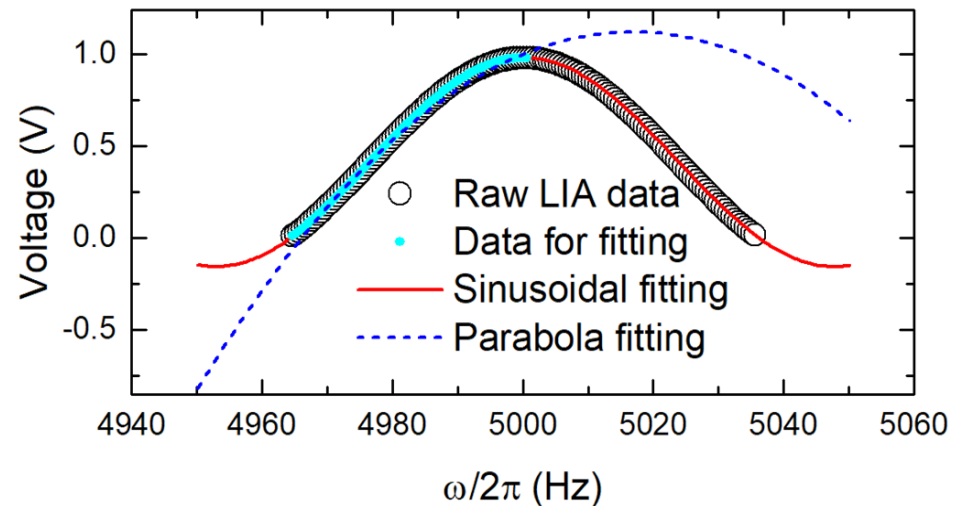
$$LIF(\Delta\omega)_{RE} = \frac{2}{T_0} \int_0^{T_0 = \frac{2\pi}{\omega_0}} \cos(\omega_0 t) \sin[(\omega_0 + \Delta\omega)t] dt = \frac{2\omega_0(\omega_0 + \Delta\omega) \sin\left(\frac{\pi\Delta\omega}{\omega_0}\right)^2}{\Delta\omega\pi(2\omega_0 + \Delta\omega)}$$

$$LIF(\Delta\omega)_{IM} = \frac{2}{T_0} \int_0^{T_0 = \frac{2\pi}{\omega_0}} \sin(\omega_0 t) \sin[(\omega_0 + \Delta\omega)t] dt = \frac{2\omega_0^2 \sin\left(\frac{\pi\Delta\omega}{\omega_0}\right) \cos\left(\frac{\pi\Delta\omega}{\omega_0}\right)}{\Delta\omega\pi(2\omega_0 + \Delta\omega)}$$

$$|LIF(\Delta\omega)| = \sqrt{LIF(\Delta\omega)_{RE}^2 + LIF(\Delta\omega)_{IM}^2}$$

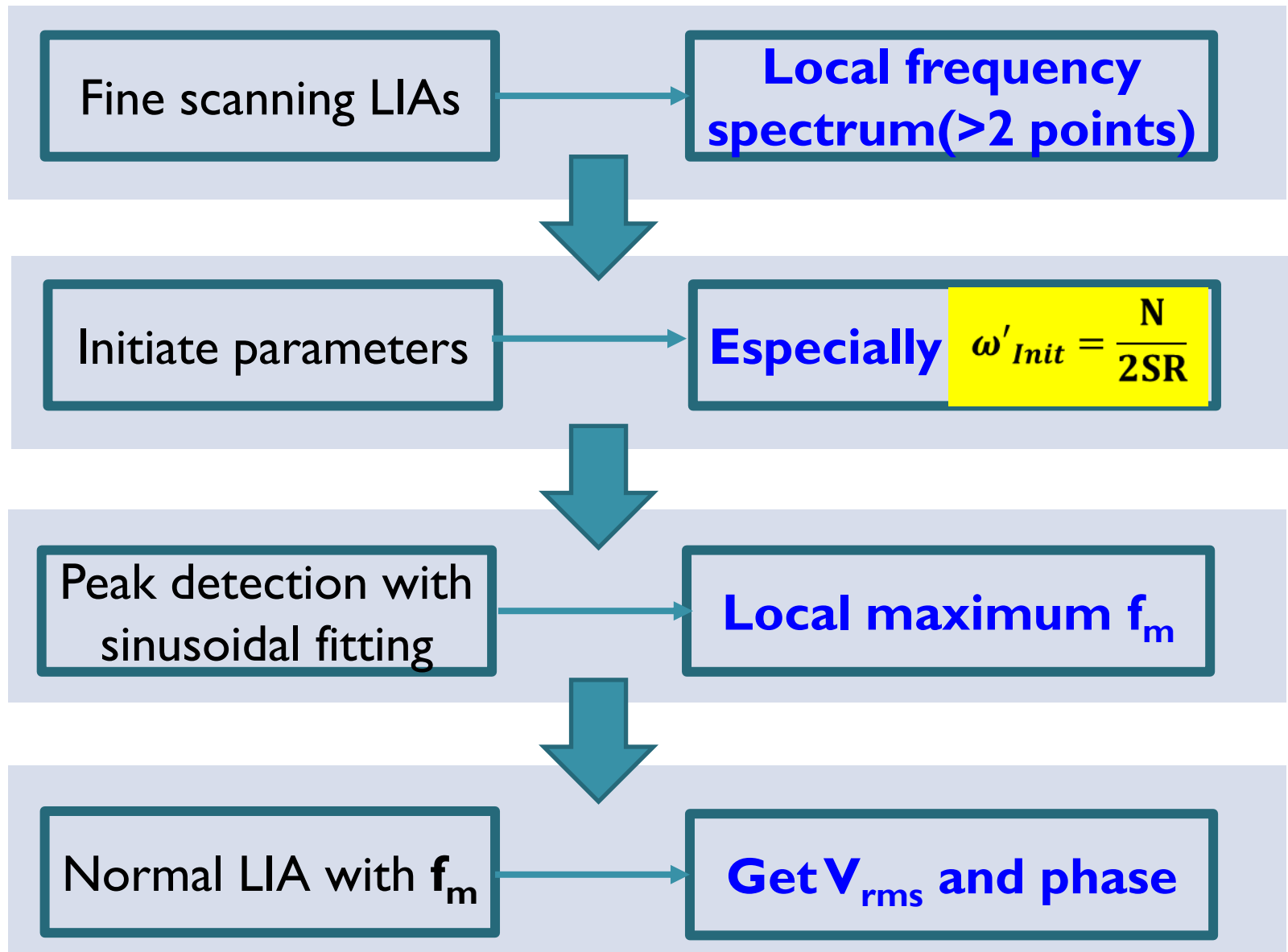
when $\Delta\omega \ll \omega_0$,

$$|LIF(\Delta\omega)| \cong \left| \frac{2\omega_0^2 \sin\left(\frac{\pi\Delta\omega}{\omega_0}\right) \cos\left(\frac{\pi\Delta\omega}{\omega_0}\right)}{\Delta\omega\pi(2\omega_0 + \Delta\omega)} \right| \approx \frac{2\omega_0 \cos\left(\frac{\pi\Delta\omega}{\omega_0}\right)}{(2\omega_0 + \Delta\omega)} \approx \cos\left(\frac{\pi\Delta\omega}{\omega_0}\right)$$

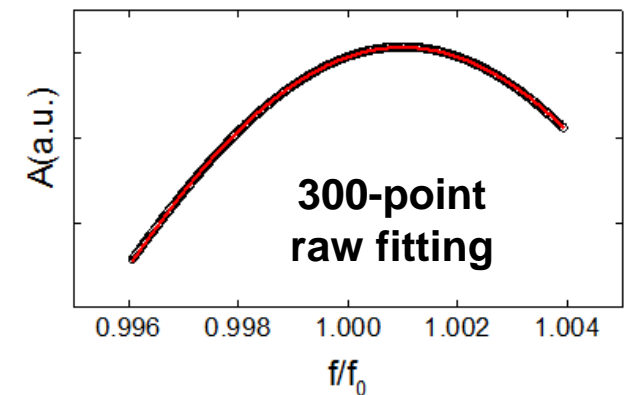
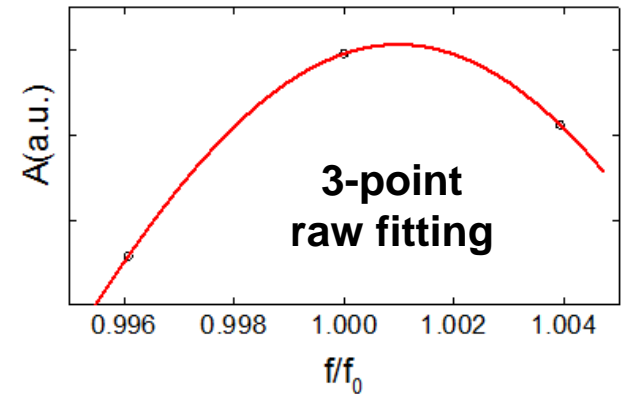
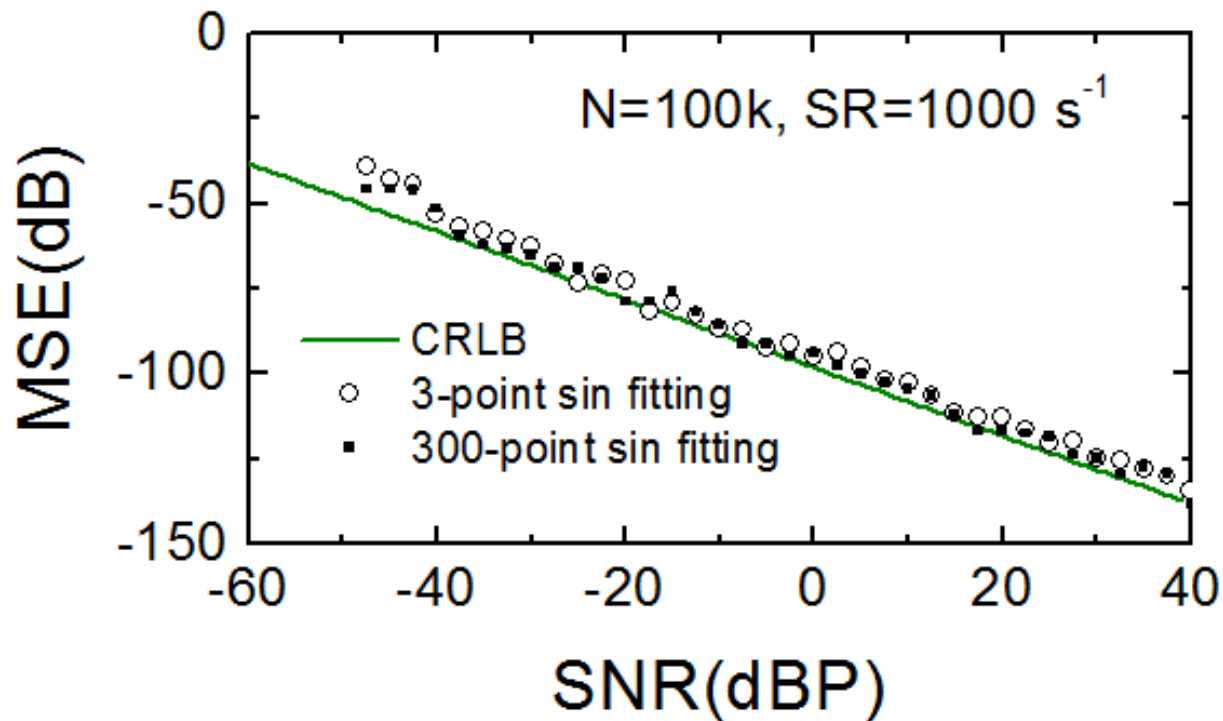


Strictly, **trigonometric function is better** than triangle and parabolic ones.

2.5 work process of LIFs

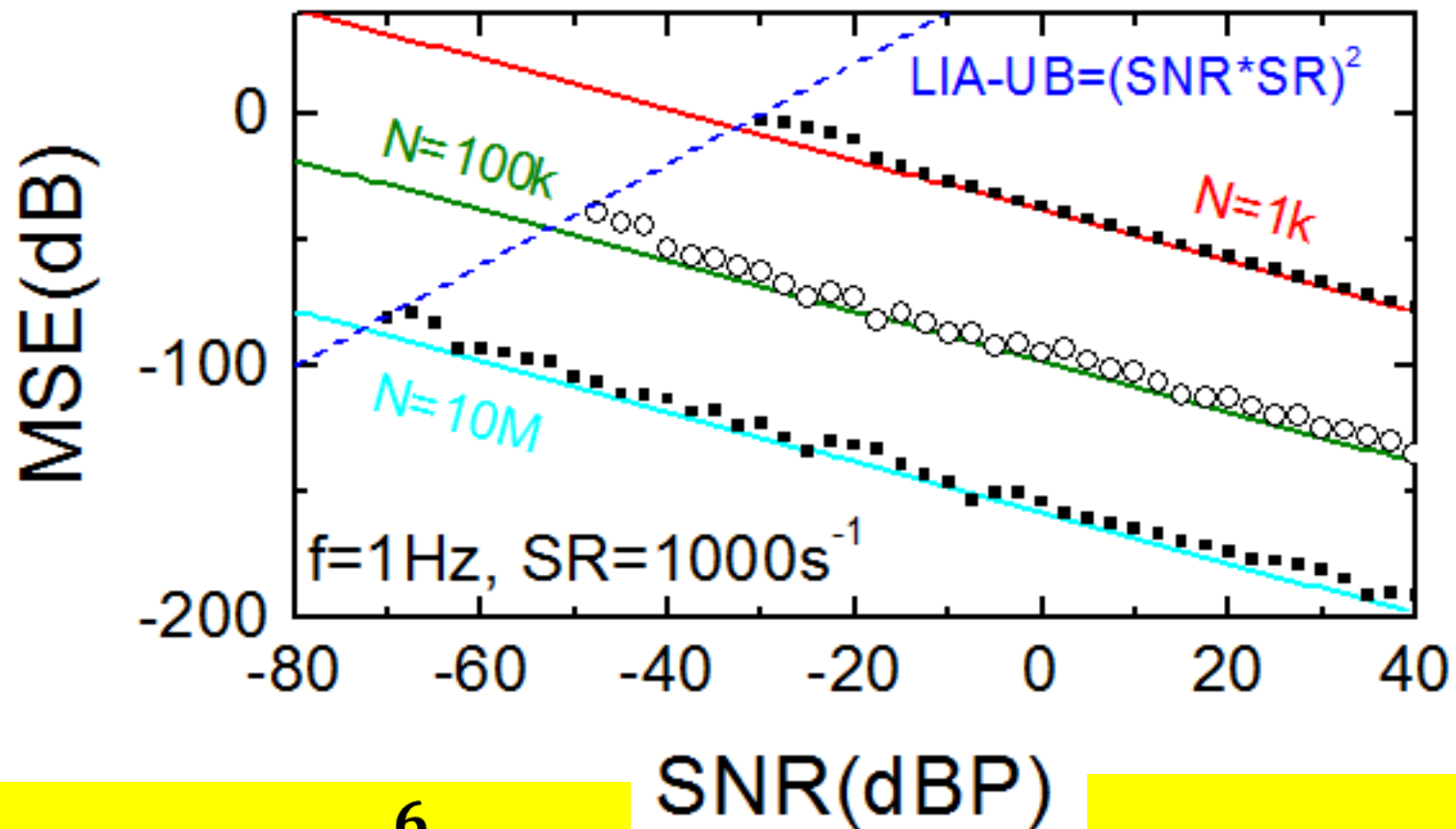


2.6 3-point sinusoidal fitting is enough



Calculation complexity in total $\sim 6N$
for one turn (multiplication and plus)

2.7 Performance simulation of LIFs(I)

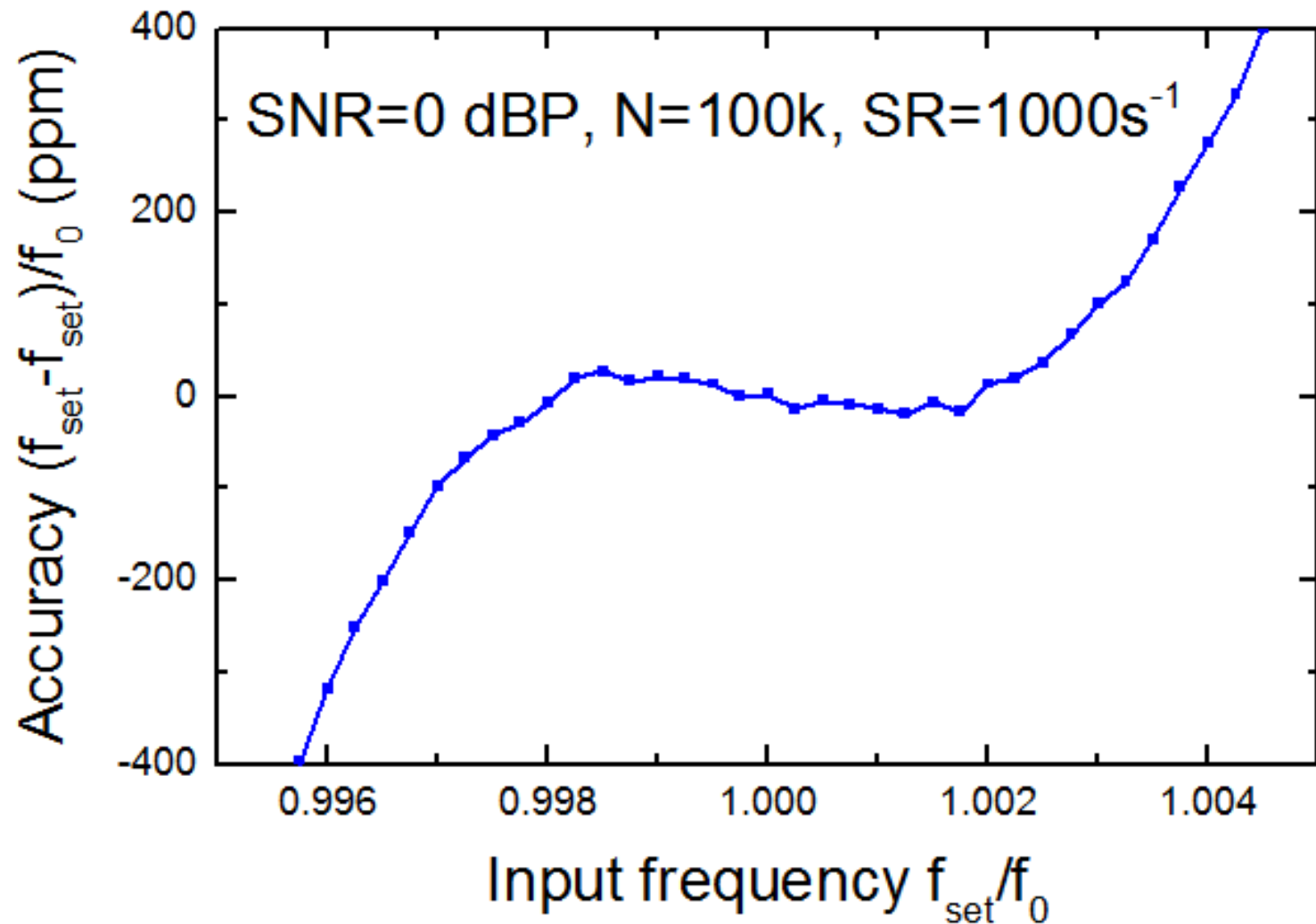


$$var(f)_{LB} \approx \frac{6}{(2\pi)^2 R_p T^2 N^3}$$

$$var(f)_{UB} \approx \frac{R_p}{T^2}$$

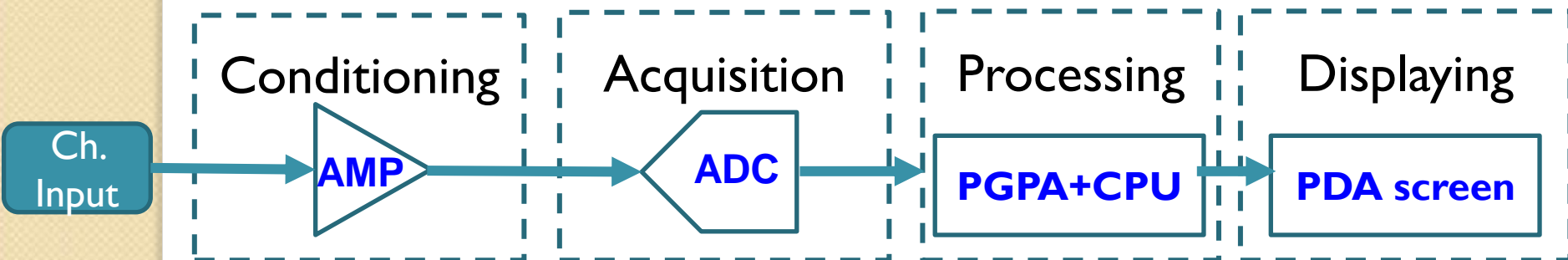
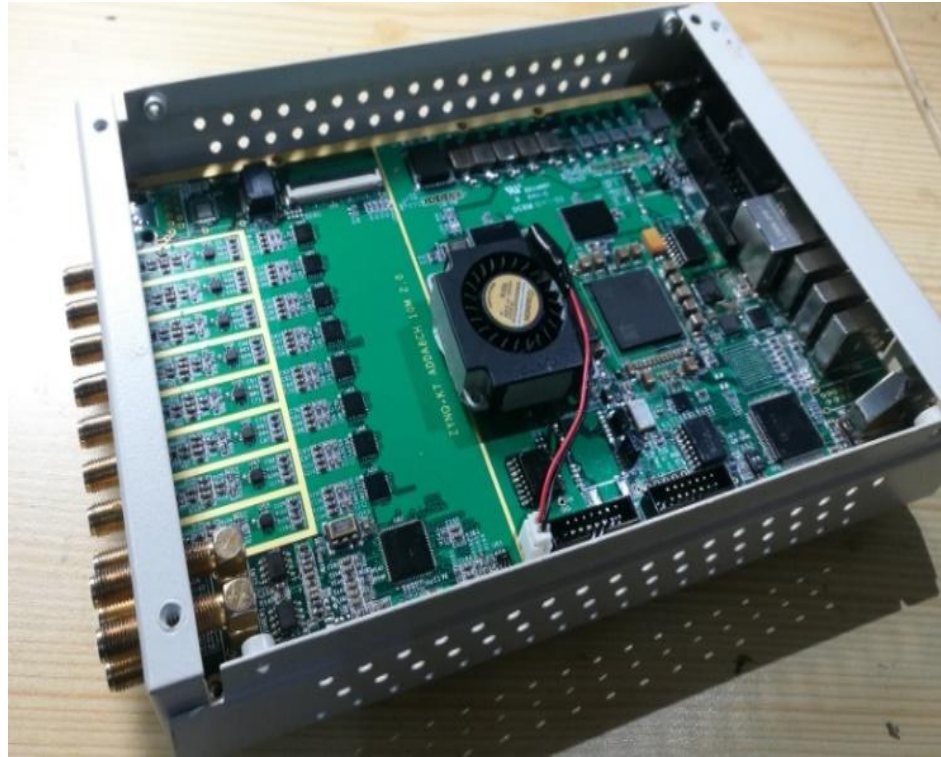
We agree to the **Cramer-Rao lower bound(CRLB)** and **LIA upper bound**

2.8 Performance simulation of LIFs(2)

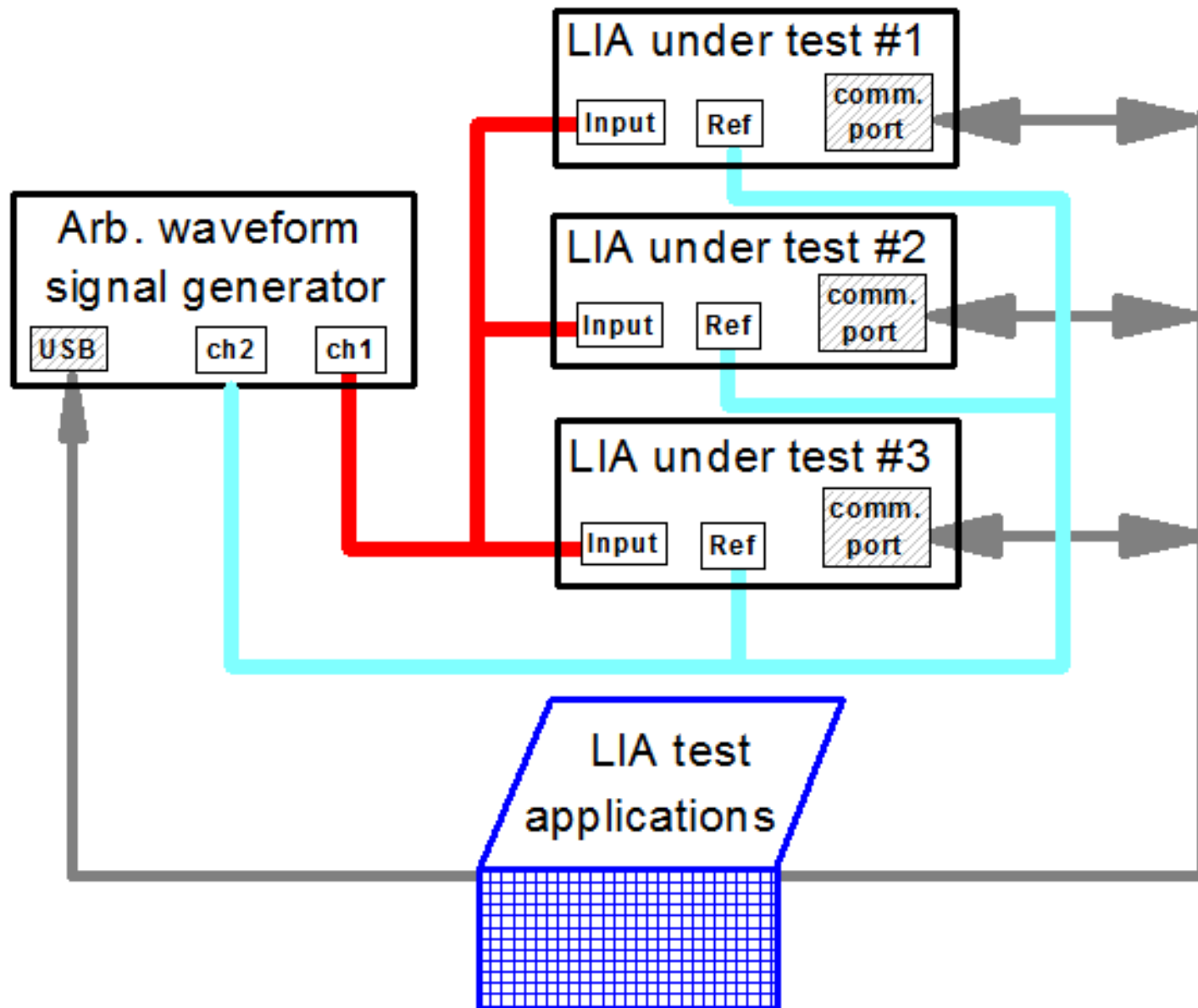


Frequency accuracy at lower SNR is reasonable high.

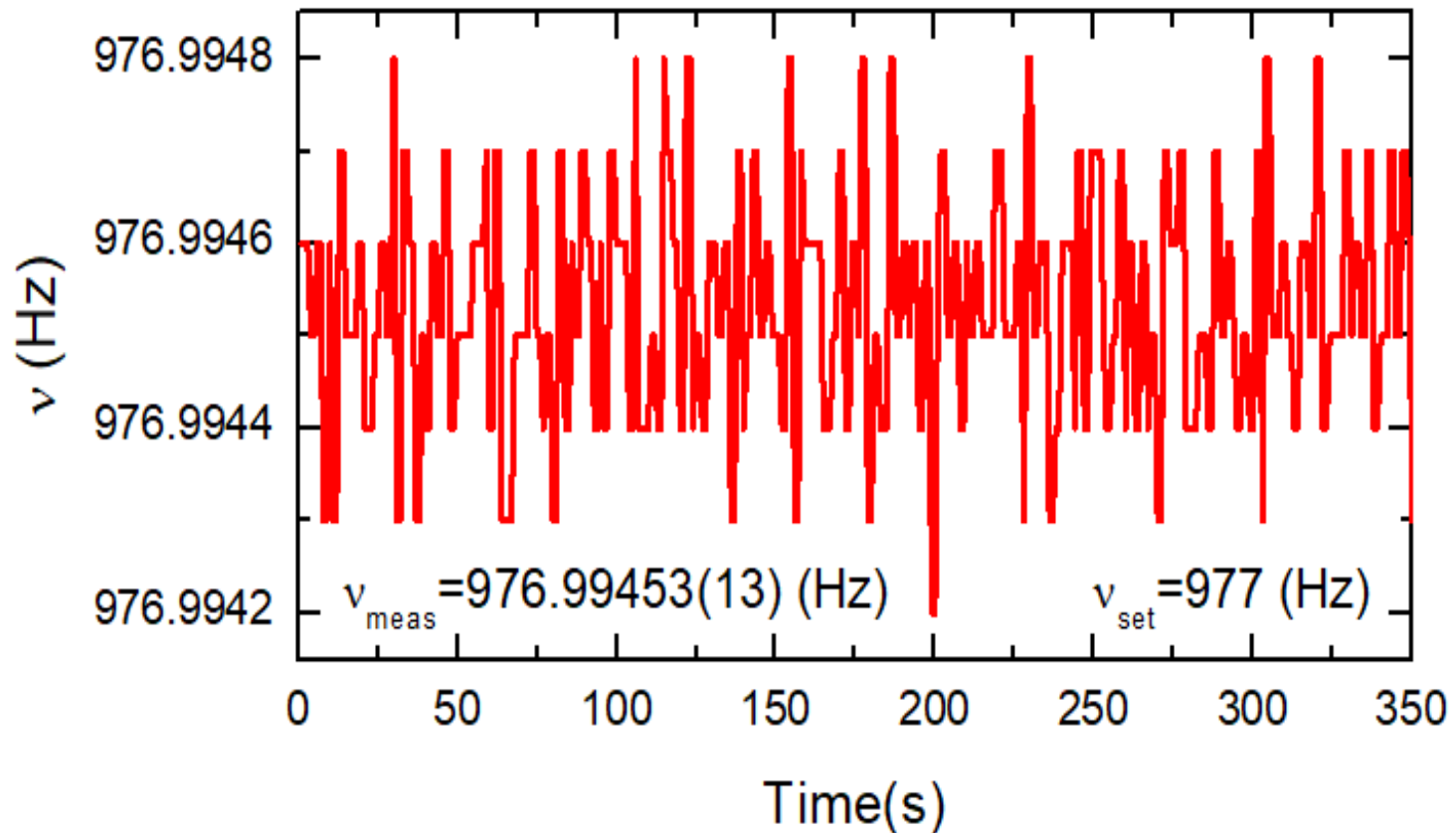
3.1 Implementation and appearance



3.2 Test setup

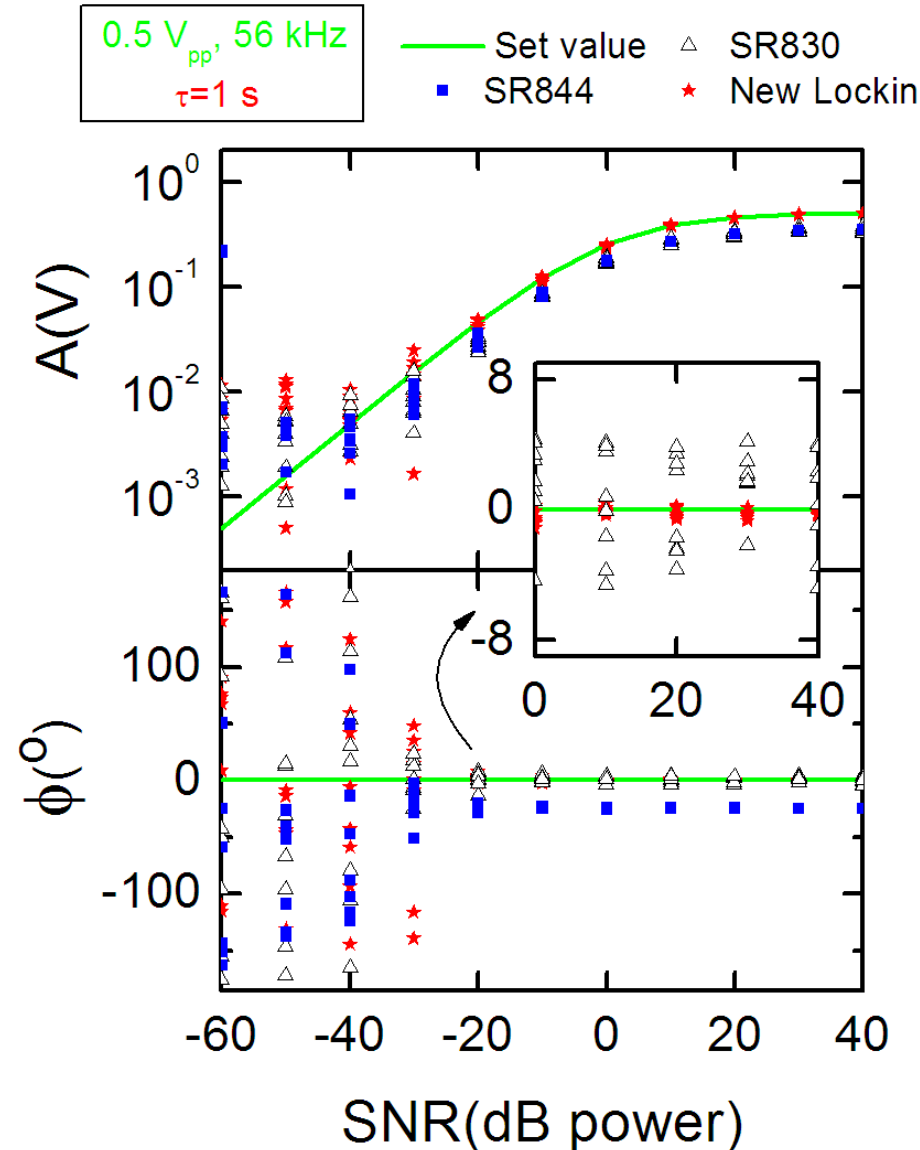
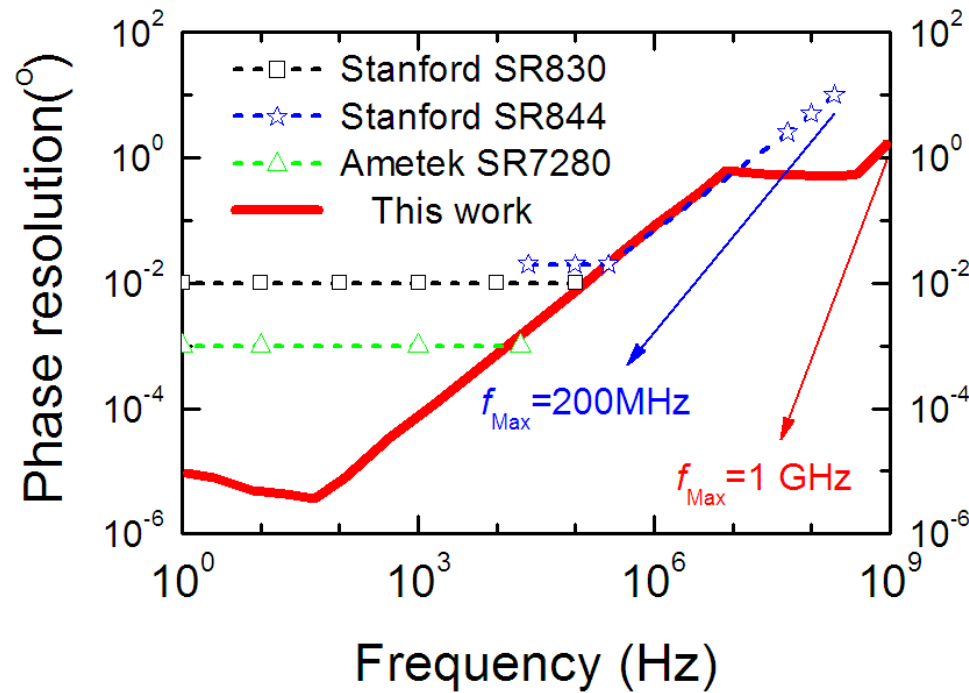


3.3 Measurements for real signal.



**Real measurement of a 977 Hz signal
with 130 micro Hz uncertainty.**

3.4 Comparison of amplitude/phase/bw



**Bandwidth, precision,
and phase resolution
is better than some
commercial LIAs**

4. Summary of Lu's LIFs

- 1、 **Principle innovation** in both frequency measurement and LIAs.
- 2、 **Better fitting function** of local frequency.
- 3、 Std. dev. Approach **CRLB with O(N)** cost.
- 4、 Precision as good as **0.2 ppm**.

$$f = \frac{N}{t}$$

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$$

Important formula at one glance

$$R_{ij} = \frac{1}{N} \sum_{k=0}^N [S_k \times S_{\text{ref},k}(f_i, \Psi_j, 1)]$$

$$\cos\left(\frac{\pi \Delta \omega}{\omega_0}\right)$$

$$\omega'_{\text{init}} = \frac{N}{2SR}$$

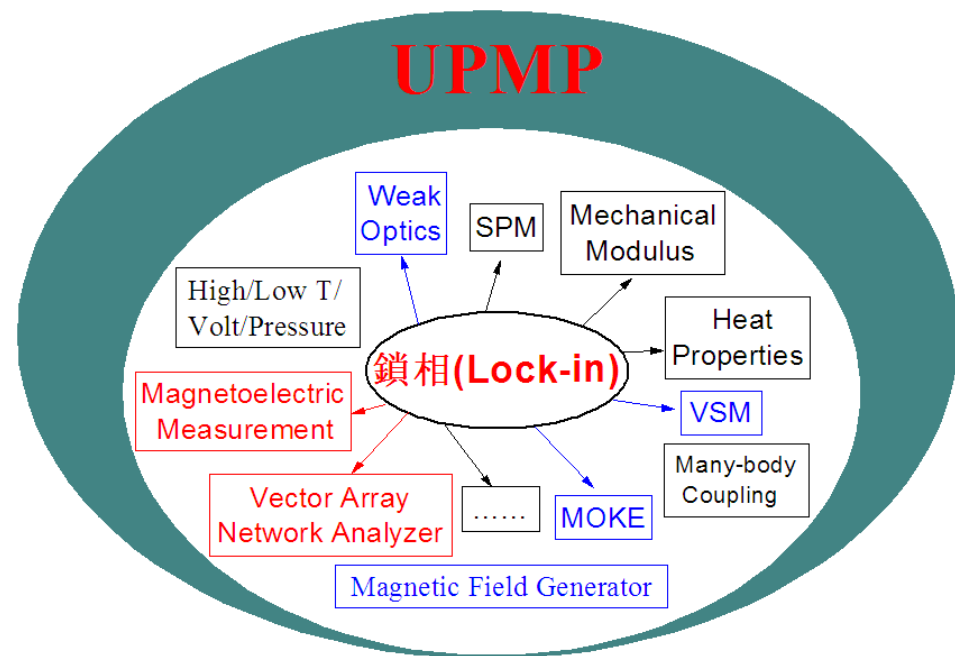
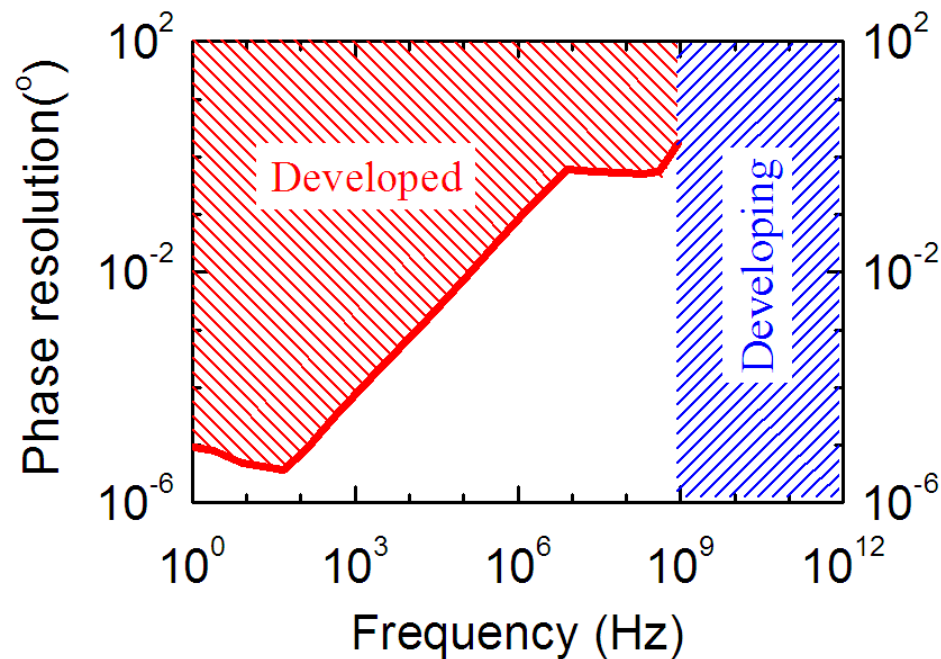
$$\text{var}(f)_{UB} \approx \frac{R_p}{T^2}$$

$$\text{var}(f)_{LB} \approx \frac{6}{(2\pi)^2 R_p T^2 N^3}$$

Acknowledgement



Thanks for your attention!



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We R&D and produce precision broadband instruments **MADE IN CHINA.**